Robust Integral Sliding Mode Control for Uncertain Impulsive Stochastic Systems with Time Delays

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Abstract—This paper studies the integral sliding mode control (ISMC) for uncertain impulsive stochastic systems with time delays. An ISMC law is proposed by designing a new integral sliding surface. Then, by constructing a piecewise time-dependent Lyapunov function, the uniformly almost surely exponentially stability conditions of the closed-loop system are established. Furthermore, an optimal design algorithm of solving control gains is formulated based on the established stable conditions and linear matrix inequalities (LMIs) theory.

Keywords—uncertain impulsive stochastic systems, integral sliding surface, solving control gains

I. INTRODUCTION

The impulse system is used to describe the phenomenon of abrupt variation of state that occurs in many practical fields, such as orbital transfer of satellite, sampled-data control systems (Benford, 2008; Naghshtabrizi, Hespanha, & Teel, 2008; Pfeiffer & Foerg, 2005), etc. Thus, the control design of the impulsive systems becomes an important and challenging task in recent years and has attracted widespread attention in control field, for example see Briat (2013), Briat and Seuret (2012a, 2012b), Chen, Ruan, and Zheng (2017), Chen, Zhang and Lu (2021), Chen and Zheng (2009) and Li, Peng, and Cao (2020). Among them, sufficient stability conditions were established in Briat and Seuret (2012a, 2012b) for linear impulsive systems by utilizing a class of looped-functional Lyapunov functions, while the authors in Briat (2013) investigated the stability analysis problem of linear aperiodic impulsive systems through the use of time-varying discontinuous Lyapunov functions. The Lyapunov stability for impulsive systems is investigated in Li et al. (2020) by using event-triggered impulsive control. The authors in Chen and Zheng (2009) studied the problems of robust stability and H^{∞} -control for uncertain impulsive systems with time delay and (Chen et al., 2017; Chen, Zhang et al., 2021) investigated the L2-gain or $L2 \times l2$ -gain control design problems for linear impulsive time-delay systems.

II. SYSTEM DESCRIPTION AND PRELIMINARIES

Consider the linear uncertain impulsive stochastic systems with time delay:

$$\begin{cases} dz(t) = [A_{i}(t)z(t) + A_{di}(t)z(t-\tau) + B_{i}(u(t) + g(t, z(t), z(t-\tau)))]dt \\ + C[D_{i}(t)z(t) + D_{di}(t)z(t-\tau)]dw(t), t \neq t_{v} \end{cases}$$

$$z(t_{v}) = E_{i}z(t_{v}^{-}), v \in N$$

$$z(t_{0} + \zeta) = \varphi(\zeta), \zeta \in [-\tau, 0]$$
(1)

where $u(t) \in \mathbb{R}^{m}$ is the control input, $z(t) \in \mathbb{R}^{n}$ is the system state, $g(t, z(t), z(t - \tau)) : \mathbb{R}_{+} \times \mathbb{R}^{n} \times \mathbb{R}^{n} \to \mathbb{R}^{m}$ is the nonlinear uncertainty. In the systems $A_{i}(t) = A_{i} + \Delta A_{i}(t)$, $A_{di}(t) = A_{di} + \Delta A_{di}(t)$, $D_{i}(t) = D_{i} + \Delta D_{i}(t)$, $D_{di}(t) = D_{di} + \Delta D_{di}(t)$, $A_{i} \in \mathbb{R}^{n \times n}$, $A_{di} \in \mathbb{R}^{n \times n}$, $C_{i} \in \mathbb{R}^{n \times m}$, $D_{i} \in \mathbb{R}^{m \times n}$, $D_{di} \in \mathbb{R}^{m \times n}$, $E_{i} \in \mathbb{R}^{n \times n}$ are real constant matrices, $B_{i} \in \mathbb{R}^{n \times m}$ is the control input matrix, $\operatorname{rank}(B_{i}) = m$ and $\operatorname{rank}(E_{i}) = n \Delta A_{i}(t) \Delta A_{di}(t)$ and $\Delta D_{i}(t) \Delta D_{di}(t)$ are the parameter uncertainties and uncertainties of stochastic perturbation, respectively. $\{t_{v}\}_{v \in \mathcal{N}_{0}} \in \mathcal{T}(\sigma_{0}, \sigma_{1})$ is the impulse time sequence, w(t) is a one-dimensional Wiener motion on the complete probability space $(\Omega, \mathcal{F}, \mathcal{F}_{t}, \mathbb{P})$.

Assumption 1[]. There exist two matrices of proper dimensions S_{1i} , S_{2i} , T_{1i} , T_{2i} such that $\begin{bmatrix} \Delta A_i(t) \quad \Delta A_{di}(t) \\ \Delta D_i(t) \quad \Delta D_{di}(t) \end{bmatrix} = \begin{bmatrix} S_{1i} \\ S_{2i} \end{bmatrix} O_i(t) \begin{bmatrix} T_{1i} & T_{2i} \end{bmatrix}$, and the unknown matrix function $O_i(t)$ satisfies $\|O_i(t)\| \le 1$.

Assumption 2 []. There are known constants α_1 , α_2 such that nonlinear uncertain function $g(t, z(t), z(t-\tau))$ satisfies $\left\|g(t, z(t), z(t-\tau))\right\| \leq \alpha_1 \left\|z(t)\right\| + \alpha_2 \left\|z(t-\tau)\right\|$

Definition 1 The zero solution of system (1) is uniformly mean-square exponentially stable over $\mathcal{T}(\sigma_0, \sigma_1)$, if there exist two positive scalars X and α such that

$$E \|z(t)\|^{2} \leq XE \|\varphi\|_{\tau}^{2} e^{-\alpha(t-t_{0})}, \quad t \geq t_{0}$$

for any initial data $\varphi \in PC_{\mathcal{F}_{0}}^{b}([-\tau, 0]; \mathbb{R}^{n})$

for any initial data

Lemma []: Given $0 < A \in \mathbb{R}^{m \times n}, B \in \mathbb{R}^{m \times n}, C \in \mathbb{R}^{p \times n}, D_1 \in \mathbb{R}^{n \times q}, D_2 \in \mathbb{R}^{m \times q}, 0 < E \in \mathbb{R}^{m \times m}_{QC} + Q_1 = Q \text{ that}$. If there exists a constant $\varepsilon > 0$ such that

$$\begin{bmatrix} -A & B^{T} \\ * & -E^{-1} \end{bmatrix} + \varepsilon^{-1} \overline{D} \overline{D}^{T} + \varepsilon \overline{C}^{T} \overline{C} < 0$$
(2)

where $\overline{D}^T = \begin{bmatrix} D_1^T & D_2^T \end{bmatrix}$ and $\overline{C}^T = \begin{bmatrix} C & 0_{p \times m} \end{bmatrix}$, then the following inequality holds

$$-A + \text{He}(D_1GC) + (B + D_2GC)^T E(B + D_2GC) < 0$$
(3)

for any matrix $G \in \mathbb{R}^{q \times p}$ satisfying $\|G\| \le 1$.

The control objective in this study is to construct a SMC strategy to guarantee that the uncertain stochastic linear impulsive system with time delay (1) is uniformly mean-square exponentially stable.

III. INTEGRAL SLIDING MODE CONTROL DESIGN

A. Sliding Mode Surface Design

(5)

It is follows from the ISMC theory that the ISMC law is chosen the following form

(4)

$$u(t) = u_1(t) + u_2(t)$$

where the role of $u_1(t)$ is stabilizing the dynamic and $u_1(t)$ is designed to reject the disturbance and maintains the dynamic. To design the sliding mode surface, a continuous sliding mode function which depend on the impulse information is proposed at first. Then, $u_1(t)$ and $u_2(t)$ is designed to assure the reachability of the sliding mode surface.

Inspired by [] and [], we design an integral sliding surface for system (1) as:

where Q is designed to satisfy that QB is nonsingular and QC = 0, $Q_1 = Q(I - J)$, and the control law $u_1(t)$ is chosen as $u_1(t) = Kx(t) + K_d x(t-\tau)$, in which K and K_d are selected such that A + BK and $A_d + BK_d$ are Hurwitz matrices, respectively. Owing to that QB is nonsingular and $\operatorname{rank}(B) = m$, there exists a matrix Z > 0 such that $Q = B^T Z$.

Because the states x(t) is discontinuous at t_{κ} , it can be seeming from the continuous of s(t) at t_{κ} that the sliding surface s(t) established for system (1) is continuous on $[0,\infty)$. Based on $v(t_{\kappa}) = 1$ and $v(t_{\kappa}) = 0$, it follows from

$$s(t_{\kappa}) = (QC + Q_1)x(t_{\kappa}^-) - \int_0^{t_{\kappa}} Q(A + BK)x(\tau)d\tau$$
$$= s(t_{\kappa}^-)$$

so, s(t) is continuous on $[0,\infty)$.

To guarantee the reachability of the designed sliding surface s(t) = 0, the SMC law $u_2(t)$ is designed in Theorem 1.

Theorem 1: Consider the system (1) under Assumptions -, if we adopt the sliding mode function () and the SMC law:

$$u_2(t) = -\phi(t)\operatorname{sign}(s(t)) \tag{7}$$

where

(6)

$$\phi(t) = \phi_0 + \left\| (QB)^{-1}QS_1 \right\| (\left\| T_1 x(t) \right\| + \left\| T_2 x(t-\tau) \right\|)$$
$$+ \alpha_1 \left\| x(t) \right\| + \alpha_2 \left\| x(t-\tau) \right\|$$

with $\phi_0 > 0$ being a scalar, then the reachability of the sliding mode surface s(t) = 0 can be ensured.

Proof: For $t \in [t_{\kappa}, t_{\kappa+1}), \kappa \in \mathcal{N}_0$, the time derivation of s(t) along (8) as follows:

$$\dot{s}(t) = Q[\Delta A(t)x(t) + \Delta A_d(t)x(t-\tau) + B(u(t) + g(t, x(t), x(t-\tau)))] - QBu_1(t)$$
(8)

Substituting (10) into (11), we get

$$\dot{s}(t) = Q[\Delta A(t)x(t) + \Delta A_d(t)x(t-\tau) +B(-\phi(t)\text{sign}(s(t)) + g(t, x(t), x(t-\tau)))]$$
(9)

Choosing the Lyapunov function $V = \frac{1}{2}s^{T}(t)(QB)^{-1}s(t)$ $s(t) = Qx(t) + \upsilon(t)Q_1x_p(t) - \int_0^t [Q(Ax(\varsigma) + A_dx(\varsigma - \tau) + Bu_1(\varsigma)) - \frac{1}{\sigma_0} \prod_{p=0}^{\infty} \frac{1}{\rho_0} M_p^{p}(\frac{1}{2})] d\varphi$ have

where $\phi_1 = (2\lambda_{\min}(QB))^{1/2}$. Since s(t) is continuous on $t \in [0,\infty)$, we obtain V(t) = 0, $\forall t \ge \tilde{T} \triangleq (2/(v_0v_1))(V(0^+))^{1/2}$, which means that s(t) = 0 for $\forall t \ge \tilde{T}$.

B. Stability of sliding motion

Once $t \ge \tilde{T}$, we get $s(t) = \dot{s}(t) = 0$ from Theorem 1. By solving $\dot{s}(t) = 0$, the equivalent control law $u_{equ}(t)$ for $u_2(t)$ can be designed as follows:

$$u_{equ}(t) = -(GB)^{-1}G[\Delta A(t)x(t) + \Delta A_d(t)x(t-\tau)] - g(t, x(t), x(t-\tau))$$
(11)

Substituting $u(t) = u_1(t) + u_{equ}(t)$ into (1) establishes the following systems

$$\begin{cases} dx(t) = [(A + BK + (I_n - \overline{B})\Delta A(t))x(t) + (A_d + BK_d + (I_n - \overline{B})\Delta A_d(t))x(t - \tau)]dt & = \sum_{l=0}^{1} \tilde{\upsilon}_l(t)u^{1-l} \\ + C[D(t)x(t) + D_d(t)x(t - \tau)]dw(t), & t \neq t_{\kappa} \\ x(t_{\kappa}) = Jx(t_{\kappa}^-), & \kappa \in N \\ x(t_0 + \zeta) = \varphi(\zeta), & \zeta \in [-\tau, 0] \\ (12) & \text{where } \quad \overline{B} = B(B^T Z B)^{-1}B^T Z \end{cases}$$
where $v_1(t) = 1 - \tilde{\upsilon}_0(t)$. Theorem 2: Consider the a given positive number A and A and

To set up the criterion of stability for system (15), the interval $[t_{\kappa}, t_{\kappa+1})$ is divided into N subintervals $\Delta_{\kappa i} \triangleq [t_{\kappa} + id_{\kappa}, t_{\kappa} + (i+1)d_{\kappa}), i \in \overline{0, N-1}$, in which $d_{\kappa} = \frac{t_{\kappa+1} - t_{\kappa}}{N}$. Then, we define:

$$\theta_{0i}(t) = \frac{t_{\kappa} + (i+1)d_{\kappa} - t}{d_{\kappa}}, t \in \Delta_{\kappa i}, i = \overline{0, N-1}, \kappa \in \mathcal{N}_0$$
(13)

Let $\begin{array}{l} \theta_{1i}(t) = 1 - \theta_{0i}(t) \\ m_0(t) : R_+ \rightarrow [0,1] \\ \text{satisfying} \end{array}$, for $\begin{array}{l} \{t_\kappa\}_{\kappa \in \mathcal{N}_0} \in \mathcal{T}(\sigma_0, \sigma_1) \\ \text{satisfying} \end{array}$, there is

$$\overline{\upsilon}(t) = \frac{m_0(t)}{\sigma_0} + \frac{m_1(t)}{\sigma_1}$$
(14)

To exploit the impulsive structure of system (15), the piecewise time-dependent Lyapunov function candidate is introduced

$$V_{2}(t, x(t)) = \zeta(t)x^{T}(t)P(t)x(t) + \int_{t-\tau}^{t} e^{-r(t-s-\tau)}x^{T}(s)Rx(s)ds$$
(15)

In (),
$$\zeta(t) = \mu^{\upsilon(t)}$$
, $P(t) = \sum_{i=0}^{N-1} P_i(t)\tilde{\chi}_i(t)$,
 $P_i(t) = \sum_{j=0}^{1} \theta_{ji}(t)P_{i+j}$, $t \in \Delta_{\kappa i}$, $P_i > 0$, $\tilde{i} \in \overline{0, N}$, $\tilde{\chi}_i(t)$ is the

characteristic function. It is obviously that V(t) is continuous)) in each impulse time intervals $(t_{\kappa}, t_{\kappa+1})$, $\kappa \in \mathcal{N}_0$.

Based on the fact that $0 \le \upsilon(t) \le 1$, there is a number $\tilde{\upsilon}_0(t) \in [0,1]$ such that

(16)

Theorem 2: Consider the sliding mode dynamics (15), for
a given positive number
$${}^{\mu}$$
, c , if there exist matrices
 $0 < P_{\tilde{i}} \in R^{n \times n}$, $\tilde{i} \in \overline{0, N}$, $0 < Z \in R^{n \times n}$ and scalars ${}^{\varepsilon_{ij} > 0}$,
 $i \in \overline{0, N}$, $j \in \overline{0, 1}$, such that :

$$\begin{bmatrix} \Psi_{ikl} & 0\\ 0 & -Z^{-1} \end{bmatrix} + \varepsilon_{ij}^{-1} \overline{E} \overline{E}^T + \varepsilon_{ij} \overline{H} \overline{H}^T < 0$$

$$J^T P_0 J \le \mu P_N \qquad (18)$$

$$B^T Z D = 0 \qquad (19)$$

where

$$\begin{split} \overline{E}^{T} &= [S_{1}^{T} P_{i+k} I_{1} S_{1}^{T}] , \quad I_{1} = [I_{n} 0_{n}] , \quad \overline{H}^{T} = [\overline{T} 0] \\ \overline{T} &= [T_{1} T_{2}] , \quad \Psi_{ikjl} = \begin{bmatrix} \Theta_{ikjl} & P_{i+k} (A_{d} + BK_{d}) \\ * & -u^{l-1}R \end{bmatrix} + \Delta_{ik} \end{split}$$

, then system (15) is uniformly mean-square exponentially stable over $\mathcal{T}(\sigma_0, \sigma_1)$.